

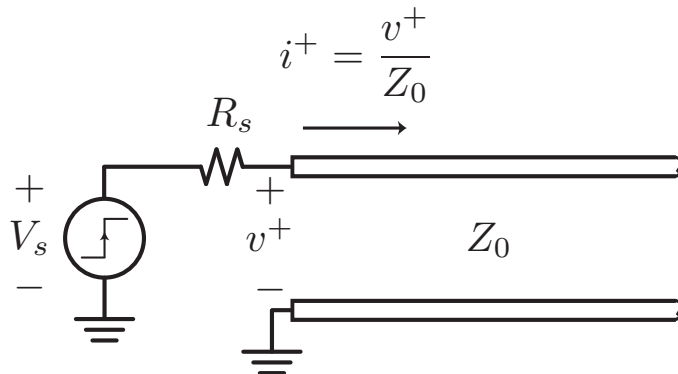
# EECS 117

## *Lecture 2: Transmission Line Discontinuities*

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# Energy to “Charge” Transmission Line



- The power flow into the line is given by

$$P_{line}^+ = i^+(0, t)v^+(0, t) = \frac{(v^+(0, t))^2}{Z_0}$$

- Or in terms of the source voltage

$$P_{line}^+ = \left( \frac{Z_0}{Z_0 + R_s} \right)^2 \frac{V_s^2}{Z_0} = \frac{Z_0}{(Z_0 + R_s)^2} V_s^2$$

# Energy Stored in Inds and Caps (I)

- But where is the power going? The line is lossless!
- Energy stored by a cap/ind is  $\frac{1}{2}CV^2/\frac{1}{2}LI^2$
- At time  $t_d$ , a length of  $\ell = vt_d$  has been “charged”:

$$\frac{1}{2}CV^2 = \frac{1}{2}\ell C' \left( \frac{Z_0}{Z_0 + R_s} \right)^2 V_s^2$$

$$\frac{1}{2}LI^2 = \frac{1}{2}\ell L' \left( \frac{V_s}{Z_0 + R_s} \right)^2$$

- The total energy is thus

$$\frac{1}{2}LI^2 + \frac{1}{2}CV^2 = \frac{1}{2} \frac{\ell V_s^2}{(Z_0 + R_s)^2} (L' + C' Z_0^2)$$

# Energy Stored (II)

- Recall that  $Z_0 = \sqrt{L'/C'}$ . The total energy stored on the line at time  $t_d = \ell/v$ :

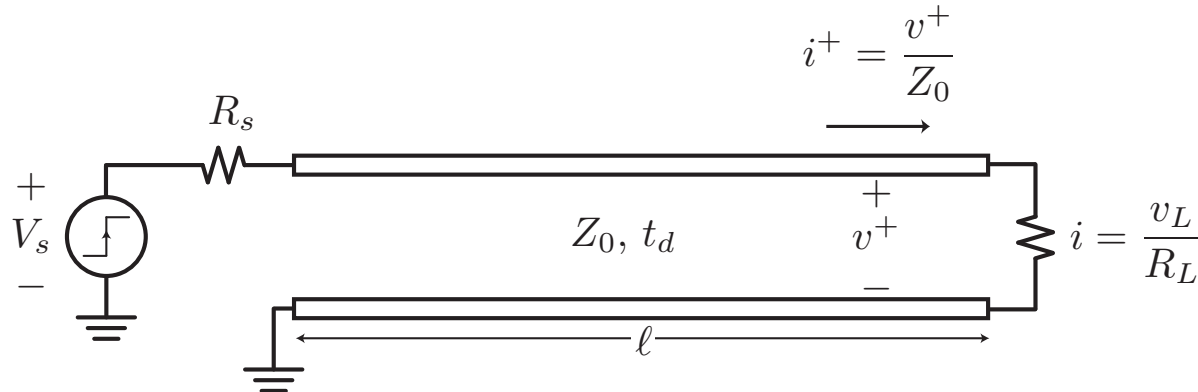
$$E_{line}(\ell/v) = \ell L' \frac{V_s^2}{(Z_0 + R_s)^2}$$

- And the power delivered onto the line in time  $t_d$ :

$$P_{line} \times \frac{\ell}{v} = \frac{\frac{\ell}{v} Z_0 V_s^2}{(Z_0 + R_s)^2} = \ell \sqrt{\frac{L'}{C'}} \sqrt{L' C'} \frac{V_s^2}{(Z_0 + R_s)^2}$$

- As expected, the results match (conservation of energy).

# Transmission Line Termination



- Consider a finite transmission line with a termination resistance
- At the load we know that Ohm's law is valid:  $I_L = V_L/R_L$
- So at time  $t = \ell/v$ , our pulse reaches the load. Since the current on the T-line is  $i^+ = v^+/Z_0 = V_s/(Z_0 + R_s)$  and the current at the load is  $V_L/R_L$ , a discontinuity is produced at the load.

# Reflections

- Thus a reflected wave is created at discontinuity

$$V_L(t) = v^+(\ell, t) + v^-(\ell, t)$$

$$I_L(t) = \frac{1}{Z_0}v^+(\ell, t) - \frac{1}{Z_0}v^-(\ell, t) = V_L(t)/R_L$$

- Solving for the forward and reflected waves

$$2v^+(\ell, t) = V_L(t)(1 + Z_0/R_L)$$

$$2v^-(\ell, t) = V_L(t)(1 - Z_0/R_L)$$

# Reflection Coefficient

- And therefore the reflection from the load is given by

$$\Gamma_L = \frac{V^-(\ell, t)}{V^+(\ell, t)} = \frac{R_L - Z_0}{R_L + Z_0}$$

- Reflection coefficient is a very important concept for transmissin lines:  $-1 \leq \Gamma_L \leq 1$
- $\Gamma_L = -1$  for  $R_L = 0$  (short)
- $\Gamma_L = +1$  for  $R_L = \infty$  (open)
- $\Gamma_L = 0$  for  $R_L = Z_0$  (match)
- Impedance match is the proper termination if we don't want any reflections

# Propagation of Reflected Wave (I)

- If  $\Gamma_L \neq 0$ , a new reflected wave travels toward the source and unless  $R_s = Z_0$ , another reflection also occurs at source!
- To see this consider the wave arriving at the source. Recall that since the wave PDE is linear, a superposition of any number of solutions is also a solution.
- At the source end the boundary condition is as follows

$$V_s - I_s R_s = v_1^+ + v_1^- + v_2^+$$

- The new term  $v_2^+$  is used to satisfy the boundary condition



# Propagation of Reflected Wave (II)

- The current continuity requires  $I_s = i_1^+ + i_1^- + i_2^+$

$$V_s = (v_1^+ - v_1^- + v_2^+) \frac{R_s}{Z_0} + v_1^+ + v_1^- + v_2^+$$

- Solve for  $v_2^+$  in terms of known terms

$$V_s = \left(1 + \frac{R_s}{Z_0}\right) (v_1^+ + v_2^+) + \left(1 - \frac{R_s}{Z_0}\right) v_1^- +$$

- But  $v_1^+ = \frac{Z_0}{R_s + Z_0} V_s$

$$V_s = \frac{R_s + Z_0}{Z_0} \frac{Z_0}{R_s + Z_0} V_s + \left(1 - \frac{R_s}{Z_0}\right) v_1^- + \left(1 + \frac{R_s}{Z_0}\right) v_2^+$$

# Propagation of Reflected Wave (III)

- So the source terms cancel out and

$$v_2^+ = \frac{R_s - Z_0}{Z_0 + R_s} v_1^- = \Gamma_s v_1^-$$

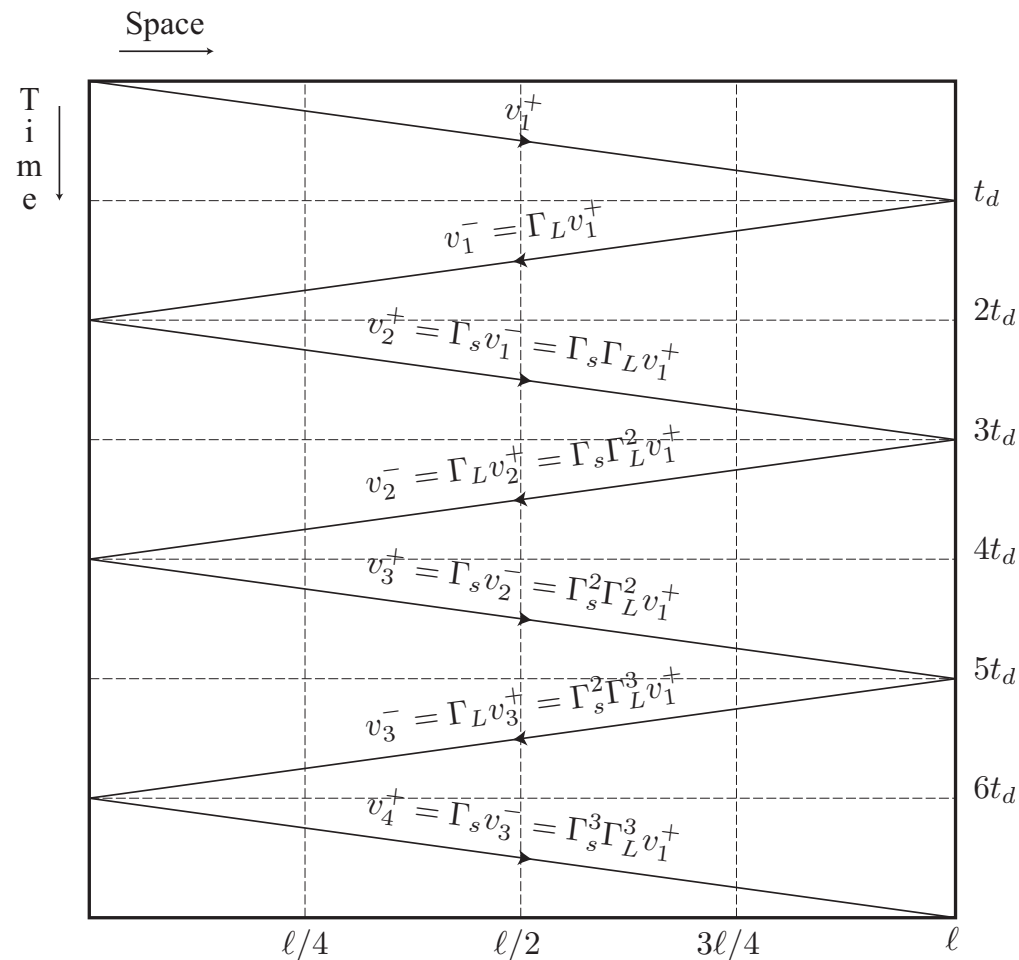
- The reflected wave bounces off the source impedance with a reflection coefficient given by the same equation as before

$$\Gamma(R) = \frac{R - Z_0}{R + Z_0}$$

- The source appears as a short for the incoming wave
- Invoke superposition! The term  $v_1^+$  took care of the source boundary condition so our new  $v_2^+$  only needed to compensate for the  $v_1^-$  wave ... the reflected wave is only a function of  $v_1^-$

# Bounce Diagram

- We can track the multiple reflections with a “bounce diagram”

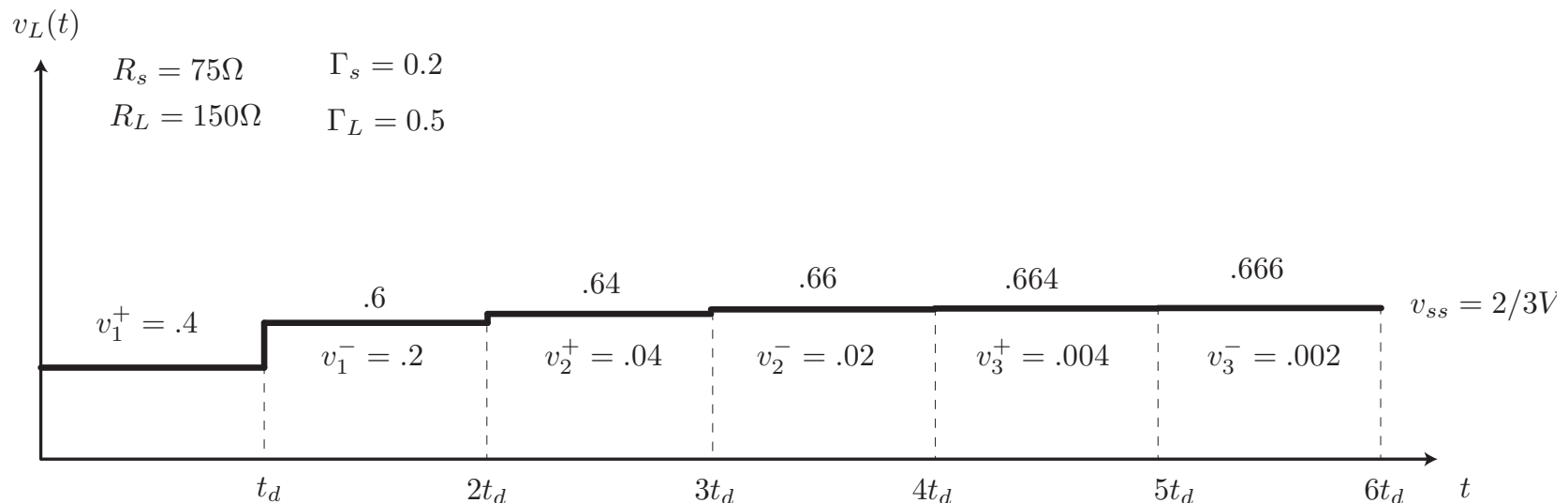


# Freeze time

- If we freeze time and look at the line, using the bounce diagram we can figure out how many reflections have occurred
- For instance, at time  $2.5t_d = 2.5\ell/v$  three waves have been excited ( $v_1^+, v_1^-, v_2^+$ ), but  $v_2^+$  has only travelled a distance of  $\ell/2$
- To the left of  $\ell/2$ , the voltage is a summation of three components:  $v = v_1^+ + v_1^- + v_2^+ = v_1^+(1 + \Gamma_L + \Gamma_L\Gamma_s)$ .
- To the right of  $\ell/2$ , the voltage has only two components:  $v = v_1^+ + v_1^- = v_1^+(1 + \Gamma_L)$ .

# Freeze Space

- We can also pick at arbitrary point on the line and plot the evolution of voltage as a function of time
  - For instance, at the load, assuming  $R_L > Z_0$  and  $R_S > Z_0$ , so that  $\Gamma_{s,L} > 0$ , the voltage at the load will increase with each new arrival of a reflection



# Steady-State Voltage on Line (I)

- To find steady-state voltage on the line, we sum over all reflected waves:

$$v_{ss} = v_1^+ + v_1^- + v_2^+ + v_2^- + v_3^+ + v_3^- + v_4^+ + v_4^- + \dots$$

- Or in terms of the first wave on the line

$$v_{ss} = v_1^+ (1 + \Gamma_L + \Gamma_L \Gamma_s + \Gamma_L^2 \Gamma_s + \Gamma_L^2 \Gamma_s^2 + \Gamma_L^3 \Gamma_s^2 + \Gamma_L^3 \Gamma_s^3 + \dots)$$

- Notice geometric sums of terms like  $\Gamma_L^k \Gamma_s^k$  and  $\Gamma_L^{k+1} \Gamma_s^k$ .  
Let  $x = \Gamma_L \Gamma_s$ :

$$v_{ss} = v_1^+ (1 + x + x^2 + \dots + \Gamma_L (1 + x + x^2 + \dots))$$

# Steady-State Voltage on Line (II)

- The sums converge since  $x < 1$

$$v_{ss} = v_1^+ \left( \frac{1}{1 - \Gamma_L \Gamma_s} + \frac{\Gamma_L}{1 - \Gamma_L \Gamma_s} \right)$$

- Or more compactly

$$v_{ss} = v_1^+ \left( \frac{1 + \Gamma_L}{1 - \Gamma_L \Gamma_s} \right)$$

- Substituting for  $\Gamma_L$  and  $\Gamma_s$  gives

$$v_{ss} = V_s \frac{R_L}{R_L + R_s}$$

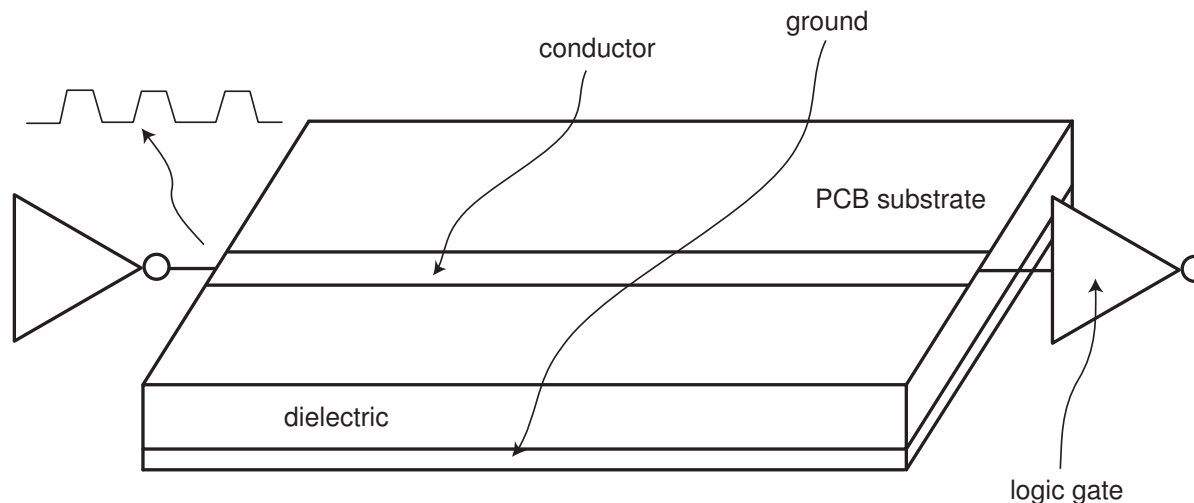
# What Happend to the T-Line?

- For steady state, the equivalent circuit shows that the transmission line has disappeared.
- This happens because if we wait long enough, the effects of propagation delay do not matter
- Conversely, if the propagation speed were infinite, then the T-line would not matter
- But the presence of the T-line will be felt if we disconnect the source or load!
- That's because the T-line *stores* reactive energy in the capaciance and inductance
- Every real circuit behaves this way! Circuit theory is an abstraction



# PCB Interconnect

- Suppose  $\ell = 3\text{cm}$ ,  $v = 3 \times 10^8\text{m/s}$ , so that  $t_p = \ell/v = 10^{-10}\text{s} = 100\text{ps}$
- On a time scale  $t < 100\text{ps}$ , the voltages on interconnect act like transmission lines!
- Fast digital circuits need to consider T-line effects



# Example: Open Line (I)

- Source impedance is  $Z_0/4$ , so  $\Gamma_s = -0.6$ , load is open so  $\Gamma_L = 1$
- As before a positive going wave is launched  $v_1^+$
- Upon reaching the load, a reflected wave of equal amplitude is generated and the load voltage overshoots  $v_L = v_1^+ + v_1^- = 1.6V$
- Note that the current reflection is negative of the voltage

$$\Gamma_i = \frac{i^-}{i^+} = -\frac{v^-}{v^+} = -\Gamma_v$$

- This means that the sum of the currents at load is zero (open)

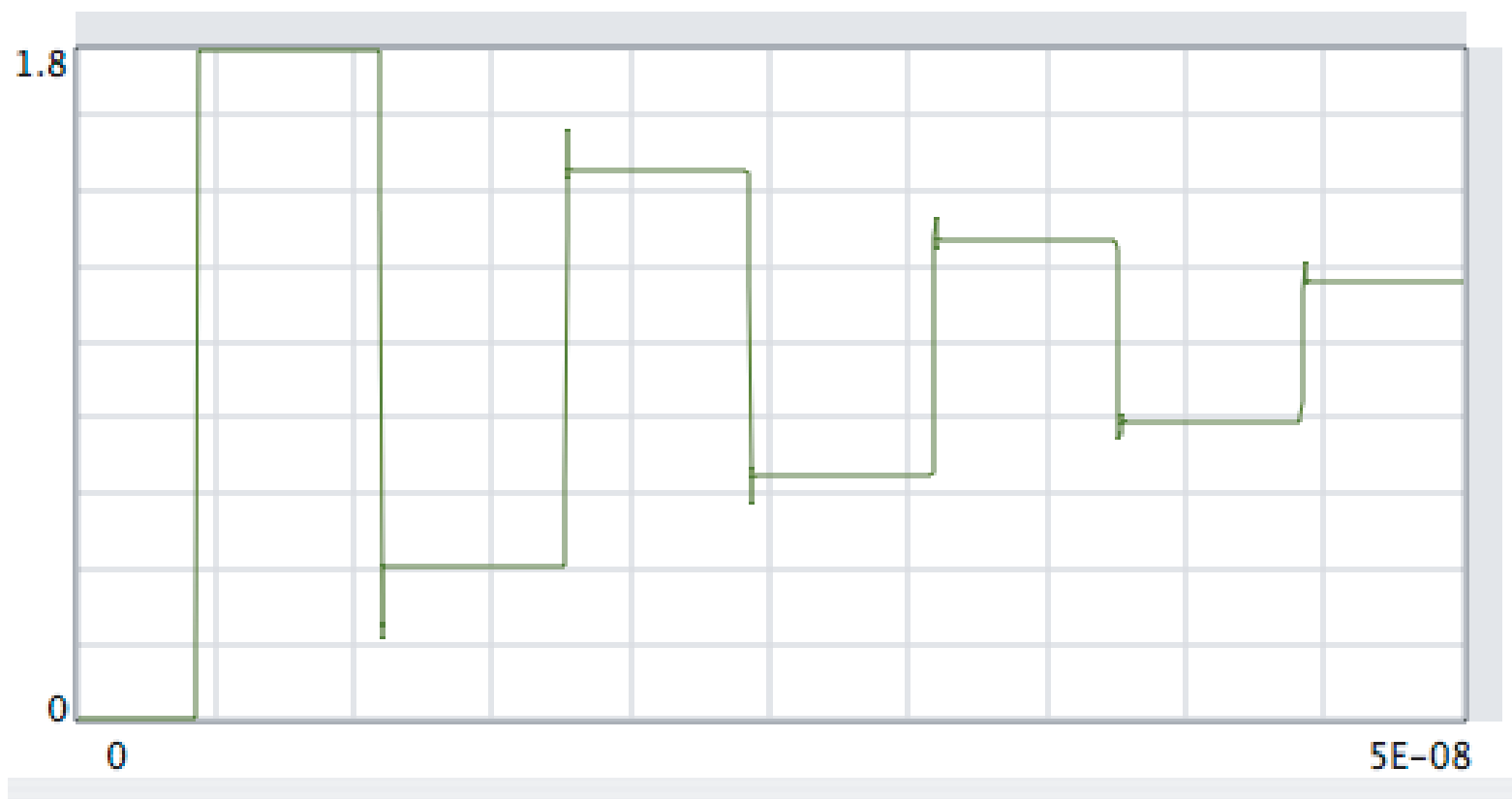
# Example: Open Line (II)

- At source a new reflection is created  $v_2^+ = \Gamma_L \Gamma_s v_1^+$ , and note  $\Gamma_s < 0$ , so  $v_2^+ = -.6 \times 0.8 = -0.48$ .
- At a time  $3t_p$ , the line charged initially to  $v_1^+ + v_1^-$  drops in value

$$v_L = v_1^+ + v_1^- + v_2^+ + v_2^- = 1.6 - 2 \times .48 = .64$$

- So the voltage on the line undershoots  $< 1$
- And on the next cycle  $5t_p$  the load voltage again overshoots
- We observe ringing with frequency  $2t_p$

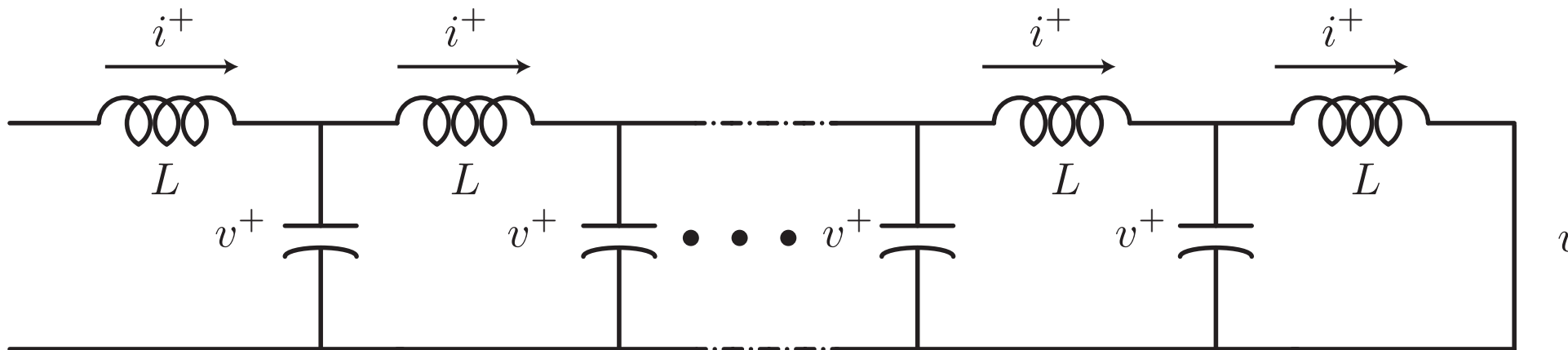
# Example: Open Line Ringing



- Observed waveform as a function of time.

# Physical Intuition: Shorted Line (I)

- The initial step charges the “first” capacitor through the “first” inductor since the line is uncharged
- There is a delay since on the rising edge of the step, the inductor is an open
- Each successive capacitor is charged by “its” inductor in a uniform fashion ... this is the forward wave  $v_1^+$



# Physical Intuition: Shorted Line (II)

- The voltage on the line goes up from left to right due to the delay in charging each inductor through the inductors
- The last inductor, though, does not have a capacitor to charge
- Thus the last inductor is discharged ... the extra charge comes by discharging the last capacitor
- As this capacitor discharges, so does its neighboring capacitor to the left
- Again there is a delay in discharging the caps due to the inductors
- This discharging represents the backward wave  $v_1^-$