

Supplementary Material for

# **Organic and Amorphous-Metal-Oxide Flexible Analogue Electronics**

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### SI 1 Impact of contact resistance on transconductance

Here we present the derivation of Equations (2.12–2.13), which provide a basic model for the impact of contact resistance on transconductance in a TFT.

The starting point is Equation (2.6), which we report here for the reader's convenience:

$$I_{DS,sat} \cong \frac{1}{2+\gamma} \mu C_I \frac{W}{L} (V_{GS} - I_{DS,sat} \cdot R_c - V_T)^{2+\gamma} (1 + \lambda V_{DS}).$$

It is useful to define here the parameter

$$K := \mu C_I \frac{W}{L} (1 + \lambda V_{DS}),$$

which allows us to rewrite the drain current in saturation as

$$I_{DS,sat} = \frac{1}{2+\gamma} \cdot K \cdot (V_{GS} - I_{DS,sat} \cdot R_c - V_T)^{2+\gamma}. \quad (\text{SI1})$$

Differentiating Equation (SI1) and arranging the terms in  $\partial I_{DS,sat} / \partial V_{GS}$  on the left-hand side leads to:

$$\begin{aligned} \frac{\partial I_{DS,sat}}{\partial V_{GS}} \cdot \left( 1 + K \cdot (V_{GS} - I_{DS,sat} \cdot R_c - V_T)^{1+\gamma} \cdot R_c \right) \\ = K \cdot (V_{GS} - I_{DS,sat} \cdot R_c - V_T)^{1+\gamma}. \end{aligned}$$

From Equation (SI1) also follows that:

$$\begin{aligned} K \cdot (V_{GS} - I_{DS,sat} \cdot R_c - V_T)^{1+\gamma} \\ = (2 + \gamma) \cdot \frac{I_{DS,sat}}{(V_{GS} - I_{DS,sat} \cdot R_c - V_T)}. \end{aligned}$$

Therefore, we have:

$$\begin{aligned} \frac{\partial I_{DS,sat}}{\partial V_{GS}} \cdot \left( 1 + (2 + \gamma) \cdot \frac{I_{DS,sat}}{(V_{GS} - I_{DS,sat} \cdot R_c - V_T)} \cdot R_c \right) \\ = (2 + \gamma) \cdot \frac{I_{DS,sat}}{(V_{GS} - I_{DS,sat} \cdot R_c - V_T)}. \end{aligned}$$

By rearranging the equation above, we obtain:

$$\begin{aligned} \frac{\partial I_{DS,sat}}{\partial V_{GS}} \cdot \left( (V_{GS} - I_{DS,sat} \cdot R_c - V_T) + (2 + \gamma) \cdot I_{DS,sat} \cdot R_c \right) \\ = (2 + \gamma) \cdot I_{DS,sat} \end{aligned} \quad (SI2)$$

$$\frac{\partial I_{DS,sat}}{\partial V_{GS}} \cdot \left( (V_{GS} - V_T) + (1 + \gamma) \cdot I_{DS,sat} \cdot R_c \right) = (2 + \gamma) \cdot I_{DS,sat}.$$

Directly from the definition of transconductance, this leads to Equation (2.12)

$$g_m = \frac{(2 + \gamma) \cdot I_{DS,sat}}{V_{GS} - V_T + (1 + \gamma) \cdot I_{DS,sat} \cdot R_c}.$$

As for Equation (2.13), it can be directly derived from Equation (2.6) (Equation (SI1)) and Equation (2.12) (Equation (SI2)). From Equation (SI1) follows that:

$$\left( \frac{1}{K} \right)^{\frac{1}{2+\gamma}} \left( (2 + \gamma) \cdot I_{DS,sat} \right)^{\frac{1}{2+\gamma}} = (V_{GS} - I_{DS,sat} \cdot R_c - V_T). \quad (SI3)$$

Plugging in Equation (SI3) into Equation (SI2), we obtain:

$$\begin{aligned} g_m \cdot \left( \left( \frac{1}{K} \right)^{\frac{1}{2+\gamma}} \left( (2 + \gamma) \cdot I_{DS,sat} \right)^{\frac{1}{2+\gamma}} + (2 + \gamma) \cdot I_{DS,sat} \cdot R_c \right) \\ = (2 + \gamma) \cdot I_{DS,sat}. \end{aligned}$$

We can rewrite this equation as

$$g_m \cdot \left( \frac{1}{K} \right)^{\frac{1}{2+\gamma}} \cdot \left( (2+\gamma) \cdot I_{DS,sat} \right)^{\frac{1}{2+\gamma}} \cdot \left( 1 + \frac{(2+\gamma) \cdot I_{DS,sat} \cdot R_c}{\left( \frac{1}{K} \right)^{\frac{1}{2+\gamma}} \cdot \left( (2+\gamma) \cdot I_{DS,sat} \right)^{\frac{1}{2+\gamma}}} \right) = (2+\gamma) \cdot I_{DS,sat},$$

from which

$$g_m \cdot \left( 1 + \frac{(2+\gamma) \cdot I_{DS,sat} \cdot R_c}{\left( \frac{1}{K} \right)^{\frac{1}{2+\gamma}} \cdot \left( (2+\gamma) \cdot I_{DS,sat} \right)^{\frac{1}{2+\gamma}}} \right) = \frac{(2+\gamma) \cdot I_{DS,sat}}{\left( \frac{1}{K} \right)^{\frac{1}{2+\gamma}} \cdot \left( (2+\gamma) \cdot I_{DS,sat} \right)^{\frac{1}{2+\gamma}}}$$

or

$$g_m \cdot \left( 1 + {}^{(2+\gamma)}\sqrt{K \cdot [(2+\gamma) \cdot I_{DS,sat}]^{1+\gamma} \cdot R_c} \right) = {}^{(2+\gamma)}\sqrt{K \cdot [(2+\gamma) \cdot I_{DS,sat}]^{1+\gamma}}. \quad (\text{SI4})$$

We now return to Equation (2.11), which gives the transconductance with zero contact resistance,  $g_{m0}$ , for a given current  $I_{DS,sat}$ . Under the hypothesis of negligible channel-length modulation, Equation (2.11) can be conveniently rewritten as:

$$g_{m0} = {}^{(2+\gamma)}\sqrt{\frac{W}{L} \cdot \mu \cdot C_I \cdot [(2+\gamma) \cdot I_{DS,sat}]^{1+\gamma}} = {}^{(2+\gamma)}\sqrt{K \cdot [(2+\gamma) \cdot I_{DS,sat}]^{1+\gamma}}. \quad (\text{SI5})$$

Comparing Equations (SI4) and (SI5), we obtain

$$g_m \cdot (1 + g_{m0} \cdot R_c) = g_{m0},$$

which is equivalent to Equation (2.13).